



In this issue we celebrate the success of the recent 2014 APSIPA Annual Summit and Conference (APSIPA ASC 2014) event which was held in Siem Reap, Cambodia, from December 9-12, 2014. It was a success in all aspects and a big number of researchers including many students congregated in this gathering to exchange knowledge and build up research networks. The opening of this newsletter issue shades lights on

some activities held in this conference. Thus, we take this opportunity to invite you all to participate in APSIPA conference in December 2015 which will be held in Hong Kong. The flyer of the conference is attached with this issue.

Three prominent keynote speakers have been invited to deliver landmarks in their research areas. Prof. P. P. Vaidyanathan from California Institute of Technology Pasadena, USA presented an interesting topic titled Ramanujan-sums and digital signal processing. He also complemented his speech by an abstract article about Ramanujan-sums, included in this newsletter issue. We encourage people to read this interesting article specially it paves the way to two extended papers in this topic published recently in IEEE Trans. on Signal Processing. The second keynote speaker

was Prof. Helen Meng from Chinese University of Hong Kong and she presented a topic titled Development of Speech Recognition Automatic and Technologies Synthesis to Support Chinese Learners of English. Professor Meng demonstrated her group approach which depends on the theory of language transfer and involves а phonological systematic comparison between the primary language and language to detect secondary mispronunciations. The third speaker was Dr. Wei-Ying Ma from Microsoft Research Asia, China. His topic was

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#	Торіс	Speaker
Tutorial 1	STRAIGHT Speech Analysis	Hideki Kawahara
Tutorial 2	Signal Processing Circuit for Pulse Transmis- sion in Solar String	Ekachai Leelarasmee
Tutorial 3	MPEG: Evolving Standards for Video Coding	Yo-Sung Ho
Tutorial 4	Signal Enhancement in Consumer Products	Akihiko Sugiyama
Tutorial 5Speaker Recognition – The Present and Fu- ture of Voice Biometrics	Speaker Recognition – The Present and Fu-	Eliathamby Ambikairajah
	and Waleed H. Abdulla	
Tutorial 6	Recent advances in Subpixel rendering for im- age/video	Oscar Au
Tutorial 7	Vector Spaces for Cross-language NLP Appli- cations	Rafael E. Banchs

titled Building a Scalable System and Algorithms for Machine Comprehension of Text. The essence of his talk was to show the current ability to learn big statistical models from large amounts of data and build comprehensive symbolic knowledge graphs from the Web.

The conference also started its activities by seven tutorials presented by leading researchers in their fields, detail of which is shown in the table above.

We encourage APSIPA community to register in our future conferences tutorials; especially they are free to attend. We also encourage them to suggest tutorials that have influence in the current research arena. ASIPA association message is to proliferate knowledge among research communities and this is why tutorials are free to attend.

Having said the rich contents of APSIPA annual conferences we encourage more participants to be involved in different activities so to achieve the goals APSIPA initiated for.

In this issue we also report on the important event IEEE SPS-APSIPA Winter School on Machine Intelligence and Signal Processing held on December 20-23, 2014. This event is the first joint activity between IEEE SPS and APSIPA which took place in New Delhi, India. It is an implementation of the Memorandum of Understanding (MoU) approved by the IEEE SPS and APSIPA in 2014 to support and sponsor joint activities.

Waleed Abdulla, APSIPA Newsletter EiC

APSIPA ASC 2014 Report

The APSIPA Annual 2014 Summit and Conference (APSIPA ASC 2014) event was held in Siem Reap, Cambodia, from December 9-12, APSIPA sixth 2014. It was the annual conference, while the previous conferences were held in Japan (2009), Singapore (2010), China (2011), USA (2012) and Taiwan (2013). Although the venue was moved from Chiang Mai due to a political situation in Thailand, APSIPA ASC 2014 had a high record of 318 accepted papers from 380 submitted papers from 24 countries. The conference had high turnouts, which were over 350 attendees.

Paper submission and presentation were organized in 6 separate tracks, Biomedical



Fig. 1 Number of accepted papers by countries

Signal Processing and Systems (BioSiPS) track, Signal Processing Systems: Design and Implementation (SPS) track, Image, Video, and Multimedia (IVM) track, Speech, Language, and Audio (SLA) track, Signal and Information Processing Theory and Methods (SIPTM) track and Wireless Communications and Networking (WCN) track, corresponding to the technical areas covered by 6 APSIPA technical committees.

The technical program included 7 tutorial sessions, 3 keynote speeches, 2 plenary overview sessions, 1 forum discussion session, together with 53 oral sessions and 7 poster sessions.

The social program included a welcome reception and a conference banquet. Siem Reap, city of Angkor Wat, is the capital city in northwestern of Cambodia, and a popular resort town as the gateway to Angkor temples region. Attendees had great experiences in the wonderful world heritage, colonial and Chinese-style architecture in the old French quarter, museums, traditional Apsara dance performances, silk



Fig. 2 Opening ceremony



Fig. 3 Keynote speeches



Fig. 4 The banquet

farms, fishing villages and a bird sanctuary near the Tonle Sap lake.

APSIPA ASC 2014 was a great success. <u>The proceedings are all accessible online</u> and can be downloaded at APSIPA website (http://apsipa.org/proceedings.htm). APSIPA ASC 2014 was also technical co-sponsored by the IEEE Signal Processing Society, and all accepted papers are accessible via IEEE Xplore as well as other Abstracting and Indexing (A&I) databases. We welcome your contributions to ASIPA ASC 2015, which will be held in Hong Kong, from December 16-19, 2015. Look forward to seeing old as well as new friends there.

APSIPA 2014 Conference Best Paper Awards

by Hsueh-Ming Hang

The Best Papers Awards in APSIPA 2014 Conference are well deserved by:

<u>IVM Track</u>: Single-sample-per-person-based Face Recognition using Fast Discriminative Multi-Manifold Analysis

Hsin-Hung Liu, Shih-Chung Hsu, and Chung-Lin Huang (*)

Department of Electrical Engineering, National Tsing-Hua University, Hsin-Chu, Taiwan

(*) Department of Applied Informatics and Multimedia, Asia University, Taichung, Taiwan

SLA Track: Modulation Spectrum-Based Post-Filter for GMM-Based Voice Conversion

Shinnosuke Takamichi, Tomoki Toda, Alan W Black (*), and Satoshi Nakamura,

Graduate School of Information Science, Nara Institute of Science and Technology (NAIST), Japan

(*) Language Technologies Institute, Carnegie Mellon University (CMU), U. S. A

SPS and BioSPS Tracks: Smoothing of Spatial Filter by Graph Fourier Transform for EEG Signals

Hiroshi Higashi, Toshihisa Tanaka (*), and Yuichi Tanaka (*)

Toyohashi University of Technology, Aichi, Japan.

(*) Tokyo University of Agriculture and Technology, Tokyo, Japan

<u>WCN and SIPTM Tracks</u>: Interference Suppression Schemes for Radio over Fiber Simultaneously Transmitted with 10 Gbps On-Off Keying

Yuya Kaneko, Takeshi Higashino, and Minoru Okada

Graduate School of Information Science, Nara Institute of Science and Technology, JAPAN









Snapshots from APSIPA 2014 Conference



Ramanujan-sums in signal processing

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Based on a keynote talk given at the APSIPA ASC, December 2014, Siem Reap, Cambodia

A great deal has been written about the Indian mathematician Srinivasa Ramanujan. Born in 1887 in Erode, Tamil Nadu, India, Ramanujan grew up with modest education. In fact he never received formal advanced training in mathematics. Nevertheless, he developed his own mathematical ideas at a very young age and made profound contributions to number theory, infinite series, and mathematical analysis [4]. His genius was noticed by a few senior mathematicians in India, who suggested that he contact some of the famous western mathematicians and communicate to them his results. Among the once that Ramanujan wrote to, the only one who took his correspondence seriously was the legendary Cambridge mathematician G. H. Hardy. Hardy was very impressed by some of the results that Ramanujan had included in his first letter to him. While some of these results were known to the western world, many were new. Hardy subsequently made arrangements for Ramanujan to visit him and work with him. This collaboration continued for about five years during which Ramanujan published many profound papers in international journals. Unfortunately he had to return to India after this due to illness, and he passed away in 1920 at the young age of 32. But the work he did during his short life span was phenomenal. Indeed, the legacy of results left behind by Ramanujan has had a great influence on mathematics, and even today mathematicians are studying some of the results he wrote in his notebooks [2], [1]. Hardy, who was himself one of the greatest English mathematicians, had the highest regard for Ramanujan. You can read about it in Hardy's own words in his introduction to a book of collected papers by Ramanujan [4].

One of the papers which Ramanujan wrote in 1918 while at Cambridge introduces a summation now known as the Ramanujan-sum [13]. Given any positive integer q the Ramanujan sum $c_q(n)$ is defined by

$$c_q(n) = \sum_{\substack{1 \le k \le q\\(k,q)=1}} e^{j2\pi kn/q}$$
(1)

where (k,q) = 1 means that k and q are **coprime**, i.e., k and q have no common divisors other than unity. Thus we add

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complex sinusoids with frequencies

$$\omega_k = \frac{2\pi k}{q} \tag{2}$$

where k and q are coprime. These are referred to as coprime frequencies with respect to the DFT grid $2\pi k/q$. Clearly $c_q(n) = c_q(n+q)$, and the Ramanujan sum can be regarded as a periodic sequence whose q-point DFT $C_q[k]$ is zero when k is not coprime to q, and nonzero (= q) when k and q are coprime. The simple-looking summation (1) has beautiful properties, and profound applications in number theory. More recently there have been interesting applications in signal processing, as we shall briefly describe below. A more elaborate account can be found in [17] and [18].

Ramanujan showed that if we add the complex numbers $e^{2\pi kn/q}$ over coprime values of k as in Eq. (1), then the result $c_q(n)$ is always real, in fact, it is **integer** valued! For example,

$$c_1(n) = 1, c_2(n) = \{1, -1\}, c_3(n) = \{2, -1, -1\}, c_4(n) = \{2, 0, -2, 0\}, c_5(n) = \{4, -1, -1, -1, -1\}, c_6(n) = \{2, 1, -1, -2, -1, 1\}, \cdots$$
(3)

and so forth. Ramanujan's motivation in introducing this sum was to show that several standard arithmetic functions in the theory of numbers can be expressed as infinite linear combinations of the form

$$x(n) = \sum_{q=1}^{\infty} \alpha_q c_q(n).$$
(4)

Just to give a flavor for the type of problems that are interesting to number theorists, consider the sum-of-divisors function $\sigma(n)$ where *n* is a positive integer. This is defined as the sum of all **divisors** (factors) of *n*, including 1 and *n*. For example $\sigma(12) = 1+2+3+4+6+12 = 28$. Ramanujan showed [13] that $\sigma(n)$ can be expanded as an infinite series

$$\sigma(n) = \frac{n\pi^2}{6} \sum_{q=1}^{\infty} \frac{c_q(n)}{q^2}, \quad n \ge 1.$$
 (5)

Many other amazing expansions for number-theoretic functions are developed in [13].

During the last decade it has been shown that the Ramanujan sum (1) has interesting applications in signal processing [9], [11], [12], [15]. For example applications in the representation of periodic signals [12], in time-frequency analysis [15], and in medicine [9] have been reported. To understand how these applications arise we have to understand some properties of Ramanujan-sums. First consider the terms

$$d_k(n) = e^{j2\pi kn/q} \tag{6}$$

in (1). Since $d_k(n) = d_k(n+q)$ it follows that this is periodic in *n*. For arbitrary *k* and *q*, the period can be either *q* or a divisor of *q* since there could be cancelations between *k* and *q*. But since *k* and *q* are restricted to be coprime in (1), the period is exactly *q*. (The period *P* of a signal x(n) is the *smallest* positive integer such that x(n) = x(n+P) for all *n*.) So each term in (1) has period *q*, although the terms have different frequencies given by (2). Now, if we add a bunch of signals with period *q* the result has period either *q* or a divisor of *q*. The beauty about the terms in (2) is that the period of the sum is exactly *q* and not smaller [17]. Thus, the Ramanujan sum $c_q(n)$ has period exactly equal to *q*.

In much of the early work, the infinite sum (4) was used as the theoretical basis and a closed form expression for α_q , developed in 1932 [3], was routinely used. Now, in signal processing applications it is convenient to consider finite duration signals $x(n), 0 \le n \le N - 1$, and obtain finite expansions instead of (4):

$$x(n) = \sum_{q=1}^{N} a_q c_q(n) \tag{7}$$

Such an expansion represents a signal as a sum of periodic components with periods $1, 2, \ldots, N$. Compare this with the DFT representation

$$x(n) = \sum_{k=0}^{N-1} b_k e^{j2\pi kn/N}$$
(8)

For each k we can write $k/N = k_i/q_i$ where q_i is a divisor of N, and $(k_i, q_i) = 1$. Thus the N terms in (8) do not have N distinct periods but only K distinct periods where K is the number of divisors of N. So, this is one of the fundamental differences between the Ramanujan expansion (7) and the DFT expansion (8): the former contains terms with all periods $1 \le q \le N$ whereas the latter contains only periods that are divisors of N. The Ramanujan-sum has therefore been found to be useful to represent a signal as a sums of periodic components, with all periods allowed. This is useful to extract hidden integer periodicities in signals, especially periods that are small compared to the data size [11], [12], [9]. It is also useful to obtain a sparse representation of signals which are superpositions of a small number of periodic components.

A more elaborate theory of signal representations using Ramanujan-sums has recently been developed in [17], [18]. First it is shown that the 1932 formula for α_q [3] is not appropriate for finite expansions (7), which are more practical. Second it is shown that the finite expansion can be done directly, by using the fact that a certain matrix based on Ramanujan sums is guaranteed to be of full rank. It is also shown that the expansion (7) leaves much to be desired. Since each period is represented by a one-dimensional subspace (i.e., just one signal $c_q(n)$), it is easy to generate examples of periodic signals such that (7) does not lead to a sparse representation. To overcome this deficiency in the traditional Ramanujan sum representations, the so-called **Ramanujan subspace** is introduced in [17], and signal representations based on Ramanujan-subspaces are developed in [18]. The idea is briefly outlined below.

We know the Ramanujan sum $c_q(n)$ is periodic in n with period q. Consider all the cyclic shifts of this sequence, namely $c_q(n-l), 0 \leq l \leq q-1$. The space spanned by these qsequences is called the Ramanujan space S_q [17]. It can be shown that this space has dimension only $\phi(q)$ where $\phi(q)$ is the number of integers¹ in $1 \leq k \leq q$ coprime to q. Furthermore, it can be shown [17] that the space S_q can be spanned by the first $\phi(q)$ cyclic shifts, that is,

$$c_q(n), c_q(n-1), \dots, c_q(n-\phi(q)+1)$$
 (9)

Any nonzero signal in the space S_q has period exactly q (it cannot be a *proper* divisor of q). Next, it is shown in [18] that any signal $x(n), 0 \le n \le N-1$ can be represented as a sum of the form

$$x(n) = \sum_{q_k|N} x_{q_k}(n) \tag{10}$$

where the notation $q_k|N$ means that q_k are divisors of N, and where $x_{q_k}(n) \in S_{q_k}$. In words, any x(n) with duration Ncan be represented as a sum of K signals $x_{q_k}(n)$ where Kis the number of divisors of N and the component $x_{q_k}(n)$ is a period- q_k signal belonging to the Ramanujan subspace S_{q_k} . This therefore leads to a decomposition of a signal into periodic components where the periods q_k are divisors of N. The relation to the DFT representation is elaborated in [18].

Finally, any two periodic components $x_{q_1}(n)$ and $x_{q_2}(n)$ in the representation (10) are orthogonal in the interval $0 \le n \le N-1$. The orthogonal periodic components $x_{q_k}(n)$ are orthogonal projetions of x(n) onto the Ramanujan subspaces S_{q_k} . It can be shown [17] that these projections can be computed using essentially **integer projection** matrices. What this means is that the computation involves only a small number of additions and subtractions of the (possibly complex valued) samples in x(n).

Since $x_{q_k}(n)$ belongs to S_{q_k} the expansion (10) is equivalent to the **double sum** representation

$$x(n) = \sum_{q_k|N} \sum_{l=0}^{\phi(q_k)-1} \beta_{kl} c_{q_k}(n-l)$$
(11)

It can be shown that $\sum_{q_k|N}\phi(q_k)=N$ so that there are exactly N terms in the summation (11). Thus

$$\{c_{q_k}(n-l)\}\tag{12}$$

forms a basis for \mathbb{C}^N , where $q_k|N$ and $0 \le l \le \phi(q_k) - 1$. A number of examples are given in [18] to demonstrate the advantages of the new Ramanujan-space representation (11) over the traditional Ramanujan representation (7), and the traditional DFT representation.

But even the representation (11) has limited usefulness because it uses only components whose periods are divisors of N (like the DFT). In order to overcome this, an even newer

 $^{{}^{1}\}phi(q)$ is called the Euler totient function [6].

representation based on the so-called **Ramanjuan frames** is developed in [16], by using a dictionary based approach. In this approach, we consider the $\phi(q)$ cyclically shifted Ramanujan sequences $c_q(n-l), 0 \le l \le \phi(q) - 1$ for all values of q in $1 \le q \le N$ (i.e., we do not restrict q to be divisors of N). Thus, the representation takes the form

$$x(n) = \sum_{q=1}^{N} \sum_{l=0}^{\phi(q)-1} y_{ql} c_q(n-l), \quad 0 \le n \le N-1$$
(13)

where N is the signal duration. The total number of terms in the right hand side is $\sum_{q=1}^{N} \phi(q) = O(N^2)$ which shows that the representation has lot of redundancy, that is, y_{ql} is not unique. But this is the beauty of the representation: we take advantage of the redundancy, and look for a solution which represents x(n) as a sum of periodic components with as small periods as possible. (Note that smaller periods imply greater structure in the signal.) An optimization problem to this effect can be formulated, by using an appropriate norm of y_{kl} in the minimization process. This is reminiscent of a sparsity constraint, but there are important differences. The details can be found in [16]. Eqn. (13) can be written in matrix form as follows:

$$\mathbf{x} = \mathbf{A}\mathbf{y} \tag{14}$$

where $\mathbf{x} = \begin{bmatrix} x(0) & x(1) & \dots & x(N-1) \end{bmatrix}^T$ and \mathbf{y} is a column vector containing the components y_{kl} in appropriate order. The **dictionary matrix A** has N rows and $\sum_{q=1}^{N} \phi(q) = O(N^2)$ columns so that it is a fat matrix. For example if N = 6 then $\sum_{q=1}^{N} \phi(q) = 12$ so that **A** is 6×12 . A unique solution \mathbf{y} can still be obtained by imposing a side constraint as explained above.

Figure 1(a) shows an example of a signal x(n), which is a superposition of three periodic signals, with periods 4, 7, and 13. The three hidden periods cannot be seen from the figure. The sum has a period equal to the least common multiple of 4, 7 and 13 which is 364. Since the signal duration is only 100, this larger periodicity cannot be seen. Figure 1(b) shows a plot of the DFT spectrum, from which it is hard to identify the hidden periods. But the Ramanujan dictionary based algorithm can readily identify the hidden periods, as shown in the Ramanujan spectrum of Fig. 1(c). There are clear peaks at 2, 4, 7, and 13. The peak at 2 arises because of the 2nd harmonic of the period 4 component. Further examples can be found in [16], [18], and [19].

A number of interesting applications and new ideas arise out of the above representations. For example if we have a signal with time varying periodic components (i.e., periodicities localized in time) then how to generalize the above techniques for **time-period localization**? And how to extend these ideas for the case of multidimensional signals with nonseparable periodic patterns (as described by generalized lattices)? These, and other questions are currently under investigation.

The great English mathematician Hardy, in good humor, had once mentioned that profound, respectable mathematics cannot have any worthwhile applications in the real world [5]. In other words, if a mathematical theory of some sort finds an useful application, then that would be "ordinary" math. He went on to give examples, but we shall omit those here. Suffice it to say that history has proved Hardy wrong. In engineering sciences of the last century, we have repeatedly found many applications of deep, profound math. Examples include coding theory, cryptography, information theory, and many areas of signal processing. The inspiring article written by Maddox in the journal *Nature* [8] is worth mentioning here. The usefulness of Ramanujan's sum in signal processing is yet another example of the fact that highly respectable math can indeed have useful applications in engineering.



Fig. 1. (a) A signal which is a superposition of three periodic components with periods 4, 7, and 13. (b) Fourier spectrum obtained from DFT, and (c) Periodicity spectrum obtained based on Ramanujan dictionary.

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THE MAN WHO KNEW INFINITY

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APSIPA Newsletter EiC

RF Fingerprinting: Research Opportunities

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Abstract—Radio Frequency (RF) fingerprinting aims to develop a unique identity for a wireless device that can be used in the same way a biological fingerprint operates, to improve the security and privacy of wireless communication. This is in contrast to the traditional bit-level algorithmic approaches to securing transmissions. In this letter, we introduce RF fingerprinting, and a comprehensive overview of research challenges and opportunities associated with its deployment in low-cost portable mobile devices.

I. INTRODUCTION

Physical layer security is a new paradigm for securing the identity of wireless devices by extracting the unique features embedded in the electromagnetic waves emitted by transmitters, called RF fingerprinting. RF fingerprinting broadly refers to the process of identifying the source of a transmission based on features extracted from its RF signal. The features of a signal can be classified as:

- a) Features specific to the channel, which describe the response of the wireless channel and its surrounding environment.
- Features specific to the transmitter, which characterize the wireless transmitter through the unique radiometric features caused by the transmitter hardware.

RF fingerprinting based on features specific to a transmitter is also called radiometric identification [1]. Radiometric identification uses only features originating from the transmitter hardware and totally ignores features of the channel, such as channel response. However, the transmitted signal passes through a wireless channel, which might change some of its attributes. In this letter, RF fingerprinting refers to the process of identifying a specific transmitter from its received signal, and no differentiation is made between features specific to the channel and features specific to the transmitter.

Figure 1 is a generic rudimentary block diagram of a digital radio communications system, which shows the transformation of a baseband signal from its origin at the transmitter to its reception at the receiver. The grey-shaded components are typical sources of transmitter fingerprints. The imperfections include modulation errors at the modulator, phase noise at oscillators, spurious tones from mixers and Power Amplifiers (PA), non-linearity distortion at PAs, power ramp distortions (which are associated with the transients), and distortion of the equivalent filter in the path from the digital module to the antenna (including the analog Intermediate Frequency (IF) filters and RF filters). Although it may be possible to eliminate these hardware imperfections through more precise



Figure 1. Radiometric block diagram showing different sources of impairments in overall digital communication system

manufacturing and quality control, this would greatly increase the cost. In fact, most common technology standards, including IEEE 802.11, IEEE 802.15.4 and IEEE 802.22, explicitly require various wireless devices to tolerate rather wide ranges of RF variations in received signals, in order to achieve seamless operation. Therefore, these transmitter-specific RF imperfections are likely to remain available for establishing the identity of a transmitter.

The system-level diagram in Figure 1 displays the symmetry of the digital radio transmitter and receiver. To a degree, the receiver can be considered a reverse implementation of the transmitter. The front ends of commodity wireless receivers are built with inexpensive analog components, which have their own imperfections and, together with channel impairments, make it difficult to extract the unique features from the received signals. Furthermore, these imperfections vary between different receivers. Hence, despite the wealth of previous work, there is a demand for the practical deployment of RF fingerprinting using today's typical low-end receivers rather than performing experiments using high-end receivers in a laboratory environment.

A typical receiver consists of an analog front-end and a digital signal processing block . The analog front-end is generally non-ideal, which includes components such as a Low Noise Amplifier (LNA), an Oscillator, a Mixer, filters, capacitors, inductors and Analog to Digital Convertors. Additionally, the front-end is built on a Printed Circuit Board (PCB), where the components are mounted and connected via tracks [2]. Therefore, the design of the PCB and variations in its different parameters (e.g tracks width, layers, component placement) can introduce interference and distort the signal.

Distortion in most analog and RF circuit blocks arises either due to inherent randomness during the manufacturing phase [3, 4] or due to component non-linearities. Non-linearities cause unwanted effects including spectral regrowth and spreading, inter-symbol interference, and constellation warping [5]. The nonlinear properties change due to variation in temperature or device aging. For example, nonlinear phenomena that are often associated with LNA are the gain compression (or expansion) and the third order intermodulation product [6]. Mixers have non-ideal properties including self-mixing and inphase/quadrature-phase (I/Q) imbalance. The nonlinear characteristics associated with oscillators include frequency offset and phase jitter. It is possible to compensate for frequency offset because it is slowly changing. The phase jitters have a strong effect in cases where the symbols are rather long, for example in OFDM systems.

Although many researchers have explored RF fingerprinting techniques, the feasibility of RF fingerprinting using today's typical portable mobile devices has not yet been successful. Researchers have not considered the limitations of the normal receiver which is built with low-cost components and has hardware imperfections embedded during the manufacturing phase. Our research work on a small number of devices has found that RF fingerprinting can be achieved with low-cost receivers for a specific transmitter-receiver pair, and that it is hard to reproduce an RF fingerprint from any other device that will deceive the RF fingerprinting system [7]. The dependency of RF fingerprinting on different parameters such as change in device temperature, components aging and wireless channel impairments are open research questions, which need to be addressed.

A. Open Research Challenges

Modeling the impairments of transceivers: Majority of the previous research into RF fingerprinting has experimentally evaluated the effect of impairments present in the front end of low-end devices. Our results reported in [7, 8] are based on the experimental testbed that consists of seven low-end transceivers, which were made of similar components. Performing experiments with a large number of receivers and transmitters would further validate our results. However, it is quite difficult to perform experiments on a large test bed owing to the cost of the equipment and the experimental time involved. Therefore, a novel theoretical framework is required, which should model the imperfections arising due to the modulator, oscillators, mixers and Power Amplifiers (PA), power ramp distortions, and distortion of the equivalent filter in the path from the digital module to the antenna and from various analog components in the transmission and reception chains. Developing a theoretical model of a transceiver, which

could model the random impairments of different components at front end would be a major contribution. The theoretical model will enable the researchers to investigate the likelihood of two receivers forming the same fingerprint of a single transmitter in order to analyze the reliability and robustness of RF.

Effect of mobility and device aging: Majority of the RF fingerprinting experiments have deployed the transmitter and receiver at a fixed location during the measurements. Therefore, variations in channel characteristics were limited. Our preliminary simulation results show that channel impairments have limited effect on the performance of RF fingerprinting [9]. Measurements should be performed for a moveable transmitter and receiver, and data captured over many days to further investigate the effect of channel impairments. This would give an insight into any variation that can occur in an RF fingerprint over time.

Device aging would deteriorate the parameters of different RF components in front-end of transmitter and receiver. Furthermore variation in the temperature of the device and environment will also have some effects on an RF fingerprint. This would change the RF fingerprint of the transmitter over the time. To the best knowledge of the authors, there is no research article available to support this argument. Therefore, the effect of device aging and temperature variations is an open research issue and requires further investigation.

To compensate for changes in the RF fingerprint that could arise over time as transmitters' age, an unsupervised fast, reliable, and computationally less expensive incremental machinelearning algorithm should be developed. The unsupervised machine-learning algorithm should enable a wireless device to autonomously create an RF fingerprint and provide secrecy and privacy of the wireless device.

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APSIPA Transactions on Signal and Information Processing is pleased to introduce a new type of paper submission that aims to facilitate the process of converting accepted conference papers into strong journal publications.

These new **Express White Papers** (EWP) will consist of an already accepted/published conference publication, together with a short summary of extensions to be developed by the authors to prepare a full journal submission. The editors will provide feedback within 2 weeks, with the goal of helping the authors prepare a stronger journal submission.

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Editor-in-Chief

Antonio Ortega (University of Southern California, USA)



IEEE SPS-APSIPA Winter School on Machine Intelligence and Signal Processing, December 20-23, 2014

The past APSIPA President, Jay Kuo, and the President of IEEE Signal Processing Society (SPS), Alex Acero, discussed the possibility of building a partnership between APSIPA and the IEEE SPS in a breakfast meeting at ICASSP (Florence, Italy) in 2014 May. The idea was overwhelmingly welcomed and an initial draft of Memorandum of Understanding (MoU) was approved by the APSIPA Board of Governors in 2014 summer. Later on, this MoU was also approved by the Executive Committee of the IEEE SPS in 2014 September. Then, an official document was signed by both Presidents in 2014 October.

There are three main areas for partnership.

SPS will be the sole IEEE party to Technically Co-sponsor APSIPA's Annual Conference, APSIPA ASC. As a by-product of that co-sponsorship, APSIPA ASC proceedings should be eligible for inclusion in the IEEE Xplore® Digital Library. There may be fees to APSIPA for such hosting, and they may change during the duration of this agreement as IEEE policies change. Hence, the decision to include annual proceedings will be made each year by APSIPA. In any year, if APSIPA does decide to include APSIPA ASC papers in the IEEE Digital Library, it will only do so with SPS as its sole IEEE sponsor.

SPS hosts seasonal schools worldwide. Those held in the Asia-Pacific region may, if the organizers choose, have APSIPA as a 50-50 financial co-sponsor.

SPS maintains a list of Distinguished Lecturers. APSIPA may invite SPS Distinguished Lecturers to any of its events with the cost covered by APSIPA.

The first joint activity was the IEEE SPS-APSIPA Winter School on Machine Intelligence and Signal Processing that took place at New Delhi, India during December 20-23, 2014 The winter school was organized by the newly established Indraprastha Institute of Information Technology (IIIT), New Delhi and was attended by forty participants, majority of them coming from academia and some from industry as well. Prof. Mrityunjoy Chakraborty, member, BOG of APSIPA, was invited by the organizers to the winter school , who attended the event and delivered a lecture on APSIPA during



the inauguration. Since APSIPA membership from India so far is rather poor, Prof. Chakraborty made special efforts to reach out to the audience explaining the benefits of getting involved with APSIPA activities. The presentation by Prof. Chakraborty included a brief description of the genesis of APSIPA and its history of evolution, its current mission & vision, organizational structure etc, and followed it up by highlighting issues like membership including e.membership, APSIPA friend labs, ATSIP, APSIPA newsletter and APSIPA ASC. The participants were urged to join APSIPA as member/ e.member and in particular join the APSIPA friend labs. Benefits of publishing in ATSIP were explained in detail. The participants were also requested to submit papers to APSIPA ASC in future and attend the ASCs, with a view to hold the ASC in India in future. Many participants showed strong interests in APSIPA and it is thus hoped that in near future, participations in APSIPA activities from India may improve considerably. More information can be found on http://misp.iiitd.ac.in/

APSIPA Members Achievements

2014 UNSW Excellence in Senior Leadership Award

The Head of School of Electrical Engineering and Telecommunications, **Prof Eliathamby Ambikairajah**, has received **the 2014 UNSW Excellence in Senior Leadership award** from the Vice-Chancellor, in recognition of his tremendous leadership of the School over the last six years, and extensive contribution to the Faculty of Engineering and to the university.

Professor Ambikairajah took over as Head fo School in 2009 and has been instrumental in taking the School to new heights and setting a very strong trajectory. This is the first time this prestigous award has been given to an EE&T Head of School.

Congratulations to Professor Ambikairajah on this well-deserved award.

See more at:

http://www.engineering.unsw.edu.au/electrical-engineering/news/2014-unsw-excellence-in-senior-leadership-award



Head of School, Prof Eliathamby Ambikairajah (L) and Vice Chancellor, Prof Fred Hilmer (R)

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Prospective authors are invited to submit either full papers, up to 10 pages in length, or short papers up to 4 pages in length, where full papers will be for the single-track oral presentation and short papers will be mostly for poster presentation.

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