Robust Demixing Filter Update Algorithm Based on Microphone-wise Coordinate Descent for Independent Deeply Learned Matrix Analysis

Naoki Makishima*, Norihiro Takamune*, Daichi Kitamura[†], Hiroshi Saruwatari*, Yu Takahashi[‡], and Kazunobu Kondo[‡] * The University of Tokyo, Tokyo, Japan

 † National Institute of Technology, Kagawa College, Kagawa, Japan

[‡] Yamaha Corporation, Shizuoka, Japan

Abstract—In this paper, we propose a robust demixing filter update algorithm for audio source separation, which is the task of recovering source signals from multichannel mixtures observed in a microphone array. Recently, independent deeply learned matrix analysis (IDLMA) has been proposed as a state-of-theart separation method. IDLMA utilizes the deep neural network (DNN) inference of source models and the blind estimation of demixing filters based on sources' independence. In conventional IDLMA, iterative projection (IP) is exploited to estimate the demixing filters. Although IP is a fast algorithm, when a specific source model is not accurate owing to an unfavorable SNR condition, the subsequent update of filters will fail. This is because IP updates the demixing filters in a sourcewise manner, where only one source model is used for each update. In this paper, we derive a new microphone-wise update algorithm that exploits all information of the source models simultaneously for each update. The microphone-wise update problem cannot be solved by IP, but instead, a new type of vectorwise coordinate descent algorithm is introduced into the proposed algorithm to realize convergenceguaranteed parameter estimation. Experimental results show that the proposed update algorithm achieves better separation performance than IP.

I. INTRODUCTION

Audio source separation aims to recover source signals from multichannel mixtures observed using a microphone array [1]. Many types of algorithm have been proposed, e.g., unsupervised (blind) methods [2]–[13] and supervised (informed) methods [14]–[17]. Independent deeply learned matrix analysis (IDLMA) [18], [19] is a state-of-the-art source separation method combining the blind estimation of a spatial model (demixing filters) and the supervised deep neural network (DNN) inference of source models. This paper also addresses the issue on improvement of IDLMA in a theoretical aspect.

In conventional separation methods including IDLMA, a computationally efficient algorithm called *iterative projection* (IP) [5], [20] is exploited to estimate the demixing filters. Although IP is a fast algorithm, when a specific source model is not accurate owing to an unfavorable SNR condition, the successive update of filters will fail hereafter. This is because IP updates the demixing filters in a sourcewise manner, where only one source model is used for each update. Therefore, development of a robust algorithm to obtain the demixing

filters is a problem requiring urgent attention.

In order to resolve the above-mentioned problem, in this paper, we derive a new *microphone-wise* update algorithm that exploits all information of the source models simultaneously for each update. The microphone-wise update problem cannot be solved by IP, but instead, a new type of vectorwise coordinate descent (VCD) algorithm is introduced into the proposed algorithm to realize convergence-guaranteed parameter estimation. Experimental results show that the proposed update algorithm achieves better separation performance than IP. The main contribution of this paper is the theoretical derivation and experimental evaluation of the new microphone-wise update algorithm. Note that further application to the DNN-based automatic selection of sourcewise and microphone-wise update algorithms is beyond the scope of this paper and is discussed in [21].

II. CONVENTIONAL METHOD

A. Formulation

We denote the numbers of microphones and sources as M and N, respectively. In this paper, we assume M = N for simplicity. The short-time Fourier transforms (STFTs) of the multichannel source, observed, and estimated signals are defined as

$$\boldsymbol{s}_{ij} = (s_{ij1}, \dots, s_{ijN}) \quad , \tag{1}$$

$$\boldsymbol{x}_{ij} = (x_{ij1}, \dots, x_{ijM}) \quad , \tag{2}$$

$$\boldsymbol{y}_{ij} = (y_{ij1}, \dots, y_{ijN}) \quad , \tag{3}$$

where i = 1, ..., I; j = 1, ..., J; n = 1, ..., N; and m = 1, ..., M are the indexes of the frequency bins, time frames, sources, and observed microphones, respectively, and

denotes the transpose. We also denote their spectrograms as $S_n \in \mathbb{C}^{I \times J}, X_n \in \mathbb{C}^{I \times J}$, and $Y_n \in \mathbb{C}^{I \times J}$, whose elements are s_{ijn}, x_{ijn} , and y_{ijn} , respectively. When the mixing system is time-invariant and the window length in the STFT is sufficiently longer than the impulse response, the following instantaneous mixing model holds:

$$\boldsymbol{x}_{ij} = \boldsymbol{A}_i \boldsymbol{s}_{ij},\tag{4}$$



Fig. 1. Overview of IDLMA.

where $A_i = (a_{i1}, \ldots, a_{iN}) \in \mathbb{C}^{I \times J}$ is the mixing matrix and a_{in} is the steering vector of the *n*th source. When A_i is a nonsingular matrix, the demixing matrix (inverse of the mixing matrix) exists and the observed signals are separated as

$$\boldsymbol{y}_{ij} = \boldsymbol{W}_i \boldsymbol{x}_{ij}, \tag{5}$$

where $\boldsymbol{W}_i = (\boldsymbol{w}_{i1}, \dots, \boldsymbol{w}_{iN})^{\mathrm{H}} \in \mathbb{C}^{I \times J} = \boldsymbol{A}_i^{-1}$ denotes the demixing matrix, $\boldsymbol{w}_{in}^{\mathrm{H}}$ denotes the demixing filter for the *n*th source, and ^H denotes the Hermitian transpose.

B. Generative Model and Cost Function

In IDLMA, the following univariate complex Gaussian distribution is assumed as a source generative model:

$$p(\mathbf{Y}_{n}) = \prod_{ij} p(y_{ij})$$
$$= \prod_{ij} \frac{1}{\pi r_{ijn}^{2}} \exp(-\frac{|y_{ijn}|^{2}}{r_{ijn}^{2}}),$$
(6)

where r_{ijn} denotes the scale parameter (source model) of the Gaussian distribution and y_{ij} is mutually independent w.r.t. *i* and *j*. We define the scale parameter matrix as $\mathbf{R}_n \in \mathbb{R}^{I \times J}$, whose elements are r_{ijn} . The marginal distribution of (6) w.r.t. *j* is super-Gaussian when the scale parameter fluctuates and is not constant w.r.t. the time frame.

The cost function of IDLMA is the negative log-likelihood of observed signals, whose minimization is equivalent to the maximization of the independence between sources. On the basis of (6), the cost function is obtained as

$$\mathcal{L}(\boldsymbol{W}) = -\log p(\boldsymbol{X})$$

= $-\log p(\boldsymbol{Y}) - J \sum_{i} \log |\det \boldsymbol{W}_{i}|^{2}$
$$\stackrel{c}{=} \sum_{i,j,n} \left[\frac{|\boldsymbol{w}_{in}^{\mathrm{H}} \boldsymbol{x}_{ij}|^{2}}{r_{ijn}^{2}} + 2\log r_{ijn} \right]$$

 $-J \sum_{i} \log |\det \boldsymbol{W}_{i}|^{2},$ (7)

where $\stackrel{c}{=}$ denotes the equality up to addition by a constant, $W = \{W_1, \dots, W_I\}$ is the set of demixing matrices, $X = \{X_1, \dots, X_M\}$ and $Y = \{Y_1, \dots, Y_N\}$ are the sets of the observed and estimated signals, respectively, and we used the variable transformation from x_{ij} to y_{ij} on the basis of (5). The aim of IDLMA is to blindly estimate W_i only from the observed mixtures with the assistance of a DNN. An overview of the separation process of IDLMA is shown in Fig. 1.

C. Row-wise Update Rule of Demixing Matrix

In [5], [20], a fast and convergence-guaranteed algorithm called IP was proposed, which can be applied to the sum of a negative log-determinant and a quadratic form. Therefore, given the source scale parameter r_{ijn} , (7) is minimized by IP w.r.t. W_i and the update rule of W_i is obtained as

$$\boldsymbol{Q}_{in} = \frac{1}{J} \sum_{j} \frac{\boldsymbol{x}_{ij} \boldsymbol{x}_{ij}^{\mathrm{H}}}{r_{ijn}^{2}},$$
(8)

$$\boldsymbol{w}_{in} = (\boldsymbol{W}_i \boldsymbol{Q}_{in})^{-1} \boldsymbol{e}_n, \qquad (9)$$

$$\boldsymbol{w}_{in} = \frac{\boldsymbol{w}_{in}}{\sqrt{\boldsymbol{w}_{in}^{\mathrm{H}} \boldsymbol{Q}_{in} \boldsymbol{w}_{in}}},\tag{10}$$

where e_n denotes the unit vector with the *n*th element equal to unity.

D. Update Rule of Scale Parameter Matrix by DNN

DNN_n is pretrained so that the scale parameter of the source signal $\tilde{S}_n \in \mathbb{C}^{I \times J}$ is predicted from an input spectrogram $|X|^{.1}$, where $X \in \mathbb{C}^{I \times J}$ is a mixture of complex-valued spectrograms in the training data and $|\cdot|^{.1}$ for matrices denotes the element-wise absolute operation. X is prepared by mixing \tilde{S}_n with a random amplitude to simulate multiple SNR conditions [19].

We denote the DNN output as $\text{DNN}_n(\cdot)$. When we define the output scale parameter matrix as $\boldsymbol{D}_n = \text{DNN}_n(|\tilde{\boldsymbol{X}}|^{\cdot 1}) \approx \boldsymbol{R}_n$, the loss function of DNN_n is defined as

$$L(\boldsymbol{D}_n) = \sum_{i,j} \frac{|\tilde{s}_{ijn}|^2 + \delta}{d_{ijn}^2 + \delta} - \log \frac{|\tilde{s}_{ijn}|^2 + \delta}{d_{ijn}^2 + \delta} - 1, \quad (11)$$

where \tilde{s}_{ijn} and d_{ijn} are the elements of \tilde{S}_n and D_n , respectively, and δ is a small value to avoid division by zero. Since minimizing (11) corresponds to a simulation for the maximum likelihood estimation of r_{ijn} in (7) (only limited to the training data), DNN_n can be approximately interpreted as an appropriate source model based on (6).



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Proposed columnwise update

Fig. 2. (a) Conventional row-wise update of W_i . (b) Proposed columnwise update of W_i .

(b)

In inference for open data, the scale parameter matrix R_n is estimated by the pretrained DNN_n as follows:

$$\boldsymbol{R}_n \leftarrow \text{DNN}_n(|\boldsymbol{Y}_n|^{.1}),$$
 (12)

$$r_{ijn} \leftarrow \max(r_{ijn}, \epsilon),$$
 (13)

where ϵ is a small value to increase the numerical stability of IP. The input of the DNN, $|Y_n|^{.1}$, is the spectrogram of each separated signal temporally obtained through the update of W_i . Thus, in IDLMA, r_{ijn} and W_i are alternatively updated by DNN and IP to output the most independent sources.

III. PROPOSED METHOD

A. Motivation

In IDLMA, the spatial model is updated by IP. Our preliminary experiments show that the separation performance of IDLMA is affected by the update order of the demixing filter in IP. This is because when the demixing filter with an inaccurate source model is updated first, the subsequent update fails. In this paper, we propose a new microphone-wise (i.e., columnwise in W_i) update algorithm of the demixing matrix that simultaneously exploits all of the source models for each update. Fig. 2 illustrates the difference between the conventional IP and the proposed columnwise update algorithm. Since the cost function w.r.t. the columnvector of W_i cannot be minimized by IP, we employ the VCD algorithm [23] to derive the update rule.

B. Cost Function w.r.t. Column Vector

We denote the column vector of W_i as $W_i = (\tilde{w}_{i1}, \ldots, \tilde{w}_{iM})$, where \tilde{w}_{im} is a microphone-wise vector although w_{in} is a sourcewise (row) vector. The cost function (7) is rewritten using \tilde{w}_{im} as follows:

$$\mathcal{L}(\boldsymbol{W})/J \stackrel{c}{=} \sum_{i} \left[\sum_{n} \boldsymbol{w}_{in}^{\mathrm{H}} \boldsymbol{Q}_{in} \boldsymbol{w}_{in} - \log |\det \boldsymbol{W}_{i}|^{2} \right]$$
$$= \sum_{i} \left[\sum_{n=1}^{N} \sum_{m_{1}=1}^{N} \sum_{m_{2}=1}^{N} w_{inm_{1}}^{*} Q_{inm_{1}m_{2}} w_{inm_{2}} - \log |\det \boldsymbol{W}_{i}|^{2} \right]$$
$$= \sum_{i} \left[\sum_{n=1}^{N} \sum_{m=1}^{N} w_{inm}^{*} Q_{inmm} w_{inm} + \sum_{n=1}^{N} \sum_{m_{1}=1}^{N} \sum_{m_{2}\neq m_{1}}^{N} w_{inm_{1}}^{*} Q_{inm_{1}m_{2}} w_{inm_{2}} \right]$$

$$-\log |\det \boldsymbol{W}_{i}|^{2} \end{bmatrix}$$

$$\stackrel{c}{=} \sum_{i} \left[\tilde{\boldsymbol{w}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im} \tilde{\boldsymbol{w}}_{im} + \tilde{\boldsymbol{w}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{h}}_{im} + \tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{w}}_{im} - \log |\det \boldsymbol{W}_{i}|^{2} \right]$$

$$\stackrel{c}{=} \sum_{i} \left[\tilde{\boldsymbol{w}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im} \tilde{\boldsymbol{w}}_{im} + \tilde{\boldsymbol{w}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{h}}_{im} + \tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{w}}_{im} - \log |\det \boldsymbol{W}_{i}|^{2} \right], \quad (14)$$

where

$$Q_{im} = \operatorname{diag}(Q_{i1mm}, \dots, Q_{iNmm}), \qquad (15)$$
$$\tilde{h}_{im} = \left(\sum_{m' \neq m} w^*_{i1m'} Q_{i1m'm}, \dots, \sum_{m' \neq m} w^*_{iNm'} Q_{iNm'm}\right). \qquad (16)$$

Here, $Q_{inm_1m_2}$ is the (m_1, m_2) th element of Q_{in} , w_{inm} is the *m*th element of w_{in} , * denotes the complex conjugate, and $B_i = (b_{i1}, \ldots, b_{iM})^{\text{H}}$ is the adjugate matrix of W_i .

C. Columnwise Update Rule of Demixing Matrix

IP cannot be applied to the minimization of the cost function (14) unlike the case of \boldsymbol{w}_{in} because (14) includes the linear terms w.r.t. $\tilde{\boldsymbol{w}}_{im}$, i.e., $\tilde{\boldsymbol{w}}_{im}^{\mathrm{H}}\tilde{\boldsymbol{h}}_{im}$ and $\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}}\tilde{\boldsymbol{w}}_{im}$. Accordingly, we derive a new coordinate descent algorithm for (14), where $\tilde{\boldsymbol{w}}_{im}$ for each m (the microphone number) is updated by finding a stationary point of the cost function w.r.t. $\tilde{\boldsymbol{w}}_{im}$ under $\tilde{\boldsymbol{w}}_{im'}$ fixed $(m' \neq m)$. Since $\boldsymbol{b}_{im}^{\mathrm{H}}$ is independent of $\tilde{\boldsymbol{w}}_{im}$, the partial derivative of (14) w.r.t. $\tilde{\boldsymbol{w}}_{im}^{*}$ is obtained as

$$\frac{1}{J}\frac{\partial \mathcal{L}(\boldsymbol{W})}{\partial \tilde{\boldsymbol{w}}_{im}^*} = \tilde{\boldsymbol{Q}}_{im}\tilde{\boldsymbol{w}}_{im} + \tilde{\boldsymbol{h}}_{im} - \frac{\boldsymbol{b}_{im}}{\tilde{\boldsymbol{w}}_{im}^{\mathrm{H}}\boldsymbol{b}_{im}}.$$
 (17)

Hereafter, we derive the update rule of \tilde{w}_{im} as in [23]. From $\partial \mathcal{L}(W) / \partial \tilde{w}_{im}^* = 0$, we obtain the following equation that satisfies the stationary-point condition:

$$\tilde{\boldsymbol{w}}_{im} = \tilde{\boldsymbol{Q}}_{im}^{-1} (\beta_{im} \boldsymbol{b}_{im} - \tilde{\boldsymbol{h}}_{im}), \qquad (18)$$

where $\beta_{im} = 1/(\tilde{w}_{im}^{\rm H} b_{im})$. From the definition of β_{im} , we have

$$\beta_{im} \tilde{\boldsymbol{w}}_{im}^{\mathrm{H}} \boldsymbol{b}_{im} - 1 = 0.$$
⁽¹⁹⁾

By substituting (18) for (19), we obtain

$$\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im} |\beta_{im}|^2 - \tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im} \beta_{im} - 1 = 0.$$
(20)

Since the first and third terms in (20) are real numbers, the second term in (20) must also be a real number, which satisfies

$$\operatorname{Im}\left[\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}}\tilde{\boldsymbol{Q}}_{im}^{-1}\boldsymbol{b}_{im}\beta_{im}\right]=0, \qquad (21)$$

where Im[·] represents the imaginary part of the variable. From $\beta_{im} \neq 0$ and (21), we have

$$\beta_{im} = \gamma_{im} (\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im})^* = \gamma_{im} \boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \tilde{\boldsymbol{h}}_{im}$$
(22)

or

$$\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im} = 0, \qquad (23)$$

where $\gamma_{im} \in \mathbb{R} \setminus \{0\}$. When (22) holds, we can derive a quadratic equation by substituting (22) into (20) as follows:

$$\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im} |\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \tilde{\boldsymbol{h}}_{im}|^2 \gamma_{im}^2 - |\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im}|^2 \gamma_{im} - 1 = 0$$
(24)

By substituting the solution γ_{im} of (24) into (22), we have

$$\beta_{im} = -\frac{\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \tilde{\boldsymbol{h}}_{im}}{2\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im}} \left(-1 \pm \sqrt{1 + \frac{\boldsymbol{b}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im}}{|\tilde{\boldsymbol{h}}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im}^{-1} \boldsymbol{b}_{im}|^2}}\right),\tag{25}$$

where the \pm sign in (25) should be positive to make $\mathcal{L}(W)$ smaller. On the other hand, when (23) holds, the solution of (20) becomes

$$\beta_{im} = \frac{e^{\mathbf{j}\phi_{im}}}{\sqrt{\boldsymbol{b}_{im}^{\mathrm{H}}\tilde{\boldsymbol{Q}}_{im}^{-1}\boldsymbol{b}_{im}}},\tag{26}$$

where $\phi_{im} \in (-\pi, \pi]$ denotes an arbitrary phase and j is the imaginary unit. Since ϕ_{im} does not change the value of $\mathcal{L}(\boldsymbol{W})$, ϕ_{im} is set to satisfy $e^{j\phi_{im}} = (\det \boldsymbol{W}_i)/|\det \boldsymbol{W}_i|$. From (18), (25), (26), and the relation $\boldsymbol{b}_{im} = (\det \boldsymbol{W}_i)^* (\boldsymbol{W}_i^{-1})^{\mathrm{H}} \boldsymbol{e}_m$, we obtain the following update rules of $\tilde{\boldsymbol{w}}_{im}$:

$$\boldsymbol{u}_{im} \leftarrow (\boldsymbol{W}_{i}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im})^{-1} \boldsymbol{e}_{m}, \tag{27}$$

$$\hat{\boldsymbol{u}}_{im} \leftarrow \boldsymbol{Q}_{im}^{-1} \boldsymbol{h}_{im},$$

$$a_{im} \leftarrow \boldsymbol{u}_{im}^{\mathrm{H}} \tilde{\boldsymbol{Q}}_{im} \boldsymbol{u}_{im},$$
(28)
(29)

$$\hat{a}_{im} \leftarrow \boldsymbol{u}_{im}^{\text{H}} \tilde{\boldsymbol{Q}}_{im} \hat{\boldsymbol{u}}_{im},$$
 (30)

$$\tilde{\boldsymbol{w}}_{im} \leftarrow \begin{cases} \frac{\boldsymbol{u}_{im}}{\sqrt{a_{im}}} - \hat{\boldsymbol{u}}_{im} & (\hat{a}_{im} = 0), \\ \frac{\hat{a}_{im}}{2a_{im}} \begin{bmatrix} 1 - \sqrt{1 + \frac{4a_{im}}{|\hat{a}_{im}|^2}} \end{bmatrix} \boldsymbol{u}_{im} - \hat{\boldsymbol{u}}_{im} & (\hat{a}_{im} \neq 0). \end{cases}$$

$$(2\omega_{im} [V]^{\omega_{im}}]$$
(31)

This update of \tilde{w}_{im} guarantees the monotonic nonincrease in $\mathcal{L}(W)$.

IV. EXPERIMENTAL EVALUATION

A. Experimental Conditions

We confirmed the validity of the proposed columnwise update by conducting a music source separation task. We compared five methods: independent low-rank matrix analysis (ILRMA) [9], DNN with Wiener filtering (DNN+WF) [24], combination of full-rank spatial covariance model and DNN source model (FSCM+DNN) [15], conventional IDLMA with row-wise IP (Row-IDLMA), and the proposed IDLMA with



Fig. 3. Recording condition of impulse responses obtained from RWCP database.

the columnwise update (Column-IDLMA). Note that ILRMA is a "blind" (unsupervised) technique, but we show its performance just for reference to understand to what extent the supervised methods (DNN+WF, FSCM+DNN, Row-IDLMA, and Column-IDLMA) can improve the performance. For all methods except DNN+WF, we updated the spatial model 500 times. For FSCM+DNN, Row-IDLMA, and Column-IDLMA, the scale parameter matrix R_n was updated by DNN_n after every 10 iterations of the spatial parameter optimization.

We used the DSD100 dataset of SiSEC2016 [25] as the dry sources and the training dataset of DNN. The 50 songs in the dev data were used to train DNN_n and the top 25 songs in alphabetical order in the test data were used for performance evaluation. The test songs were trimmed only in the interval of 30 to 60 s. To simulate reverberant mixtures, we produced two-channel observed signals by convoluting the impulse response E2A ($T_{60} = 300 \text{ ms}$) obtained from the RWCP database [26] with each source, and mixtures of bass (Ba.) and vocal (Vo.) or drums (Dr.) and Vo. were created. The recording condition of E2A is shown in Fig. 3. All the signals were downsampled to 8 kHz. An STFT was performed using a 512-ms-long Hamming window with a 256-ms-long shift. We used the signal-to-distortion ratio (SDR) [27] to evaluate the total separation performance.

In this paper, the number of hidden layers in the constructed fully connected DNN was set to four. Each layer had 1024 units, and a rectified linear unit was used for the output of each layer. To optimize the DNN, we added the term $(\lambda/2) \sum_q g_q^2$ to (11) for regularization, where g_q is the weight coefficient in DNN, and ADADELTA [28] with a 128-size minibatch was performed for 2000 epochs. The parameter ϵ was experimentally optimized and set to $(0.1 \times (IJ)^{-1} \sum_{i,j} \sigma_{ijn}^2)^{\frac{1}{2}}$. The other parameters were set to $\delta = 10^{-5}$ and $\lambda = 10^{-5}$.

B. Results

Figs. 4 and 5 show the average SDR improvements for Ba./Vo. and Dr./Vo. separation, respectively. The proposed Column-IDLMA achieves the best SDR improvement among all the state-of-the-art blind and supervised methods in both Ba./Vo. and Dr./Vo separation. In particular, Column-IDLMA outperforms Row-IDLMA by over 0.8 dB in Ba./Vo. These results confirm that the proposed columnwise update algorithm is more appropriate than IP in IDLMA, where the accuracy of the scale parameter estimation by DNN will depend on the type of source.



Fig. 4. Average SDR improvement of 25 Ba./Vo. songs.



Fig. 5. Average SDR improvement of 25 Dr./Vo. songs.

V. CONCLUSIONS

In this paper, we derived a new columnwise update algorithm of the demixing matrix in IDLMA, which simultaneously utilizes all source models for each update. Owing to this property, the proposed update algorithm is robust against the variance of the accuracy in the DNN inference w.r.t. the type of source. For this purpose, we employed VCD, which is a convergence-guaranteed algorithm applicable to the sum of the quadric, linear, and negative log-determinant terms. Experimental results showed that the proposed columnwise update algorithm has superior separation performance to the conventional row-wise update algorithm.

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